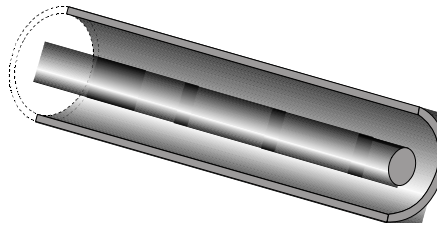


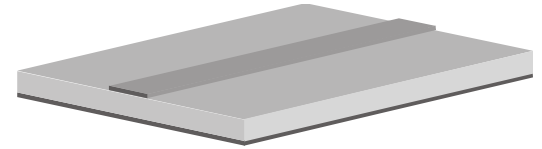
Propagazione guidata



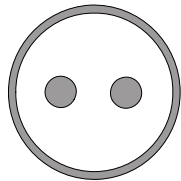
linea bifilare



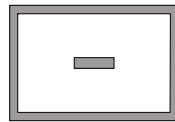
cavo coassiale



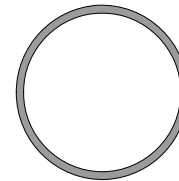
linea in microstriscia



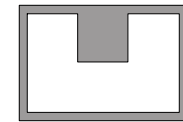
linea bifilare
schermata



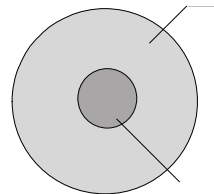
stripline



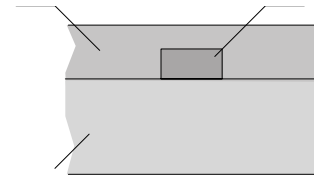
guida circolare



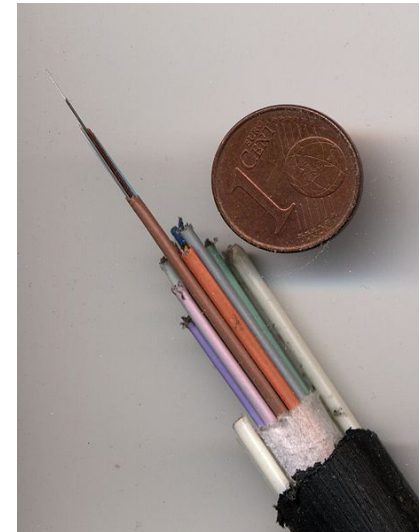
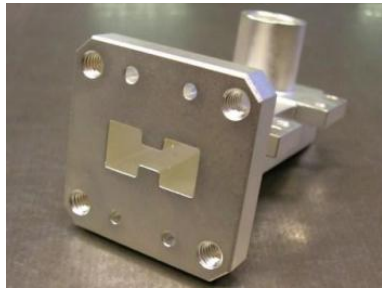
guida rettangolare
ribassata (ridge)



fibra ottica a
salto d'indice



fibra ottica integrata



Potenziali di Hertz-Debye

$$\vec{E} = \frac{\nabla \times \nabla \times \vec{u} \Psi'}{j\omega\epsilon} - \nabla \times \vec{u} \Psi''$$

$$\vec{H} = \nabla \times \vec{u} \Psi' + \frac{\nabla \times \nabla \times \vec{u} \Psi''}{j\omega\mu}$$

$$\nabla^2 \Psi + k^2 \Psi = 0$$

$$k^2 = \omega^2 \epsilon \mu.$$

- condizione di parete elettrica sui conduttori
- continuità delle componenti tangenziali dei campi sulle interfacce con altri dielettrici
- opportune condizioni all'infinito, nel caso di strutture non schermate

$$\Psi' = \Phi'(\vec{\rho}) I(z)$$

$$\Psi'' = \Phi''(\vec{\rho}) V(z)$$

$$\vec{E}_T = \frac{\eta}{jk} \nabla_T \Phi'(\vec{\rho}) \frac{\partial I(z)}{\partial z} + \vec{u}_z \times \nabla_T \Phi''(\vec{\rho}) V(z)$$

$$\vec{H}_T = -\vec{u}_z \times \nabla_T \Phi'(\vec{\rho}) I(z) + \frac{1}{jk\eta} \nabla_T \Phi''(\vec{\rho}) \frac{\partial V(z)}{\partial z}$$

$$E_z = -\frac{\eta}{jk} \nabla_T^2 \Phi'(\vec{\rho}) I(z)$$

$$H_z = -\frac{1}{jk\eta} \nabla_T^2 \Phi''(\vec{\rho}) V(z)$$

$$\frac{\partial V(z)}{\partial z} = -\gamma Z I(z)$$

$$\nabla_T^2 \Phi(\vec{\rho}) + \kappa^2 \Phi(\vec{\rho}) = 0$$

$$\frac{\partial I(z)}{\partial z} = -\gamma \frac{V(z)}{Z}$$

$$\kappa^2 = k^2 + \gamma^2$$

$$\Psi' = \Phi'(\vec{\rho}) I(z)$$

$$\Psi'' = \Phi''(\vec{\rho}) V(z)$$

$$\frac{\partial I(z)}{\partial z} = -\gamma \frac{V(z)}{Z} \quad \frac{\partial V(z)}{\partial z} = -\gamma Z I(z)$$

$$\vec{E}_T = \left(-\frac{\eta \gamma}{jkZ} \nabla_T \Phi'(\vec{\rho}) + \vec{u}_z \times \nabla_T \Phi''(\vec{\rho}) \right) V(z)$$

$$\vec{H}_T = \left(-\frac{\gamma Z}{jk\eta} \nabla_T \Phi''(\vec{\rho}) - \vec{u}_z \times \nabla_T \Phi'(\vec{\rho}) \right) I(z)$$

$$E_z = \frac{\eta \kappa^2}{jk} \Phi'(\vec{\rho}) I(z)$$

$$H_z = \frac{\kappa^2}{jk\eta} \Phi''(\vec{\rho}) V(z)$$

$$\nabla_T^2 \Phi(\vec{\rho}) + \kappa^2 \Phi(\vec{\rho}) = 0$$

$$\kappa^2 = k^2 + \gamma^2$$

guide d'onda

modi TM		modi TE	
$\nabla_{\Gamma}^2 \Phi'_i + \kappa_i'^2 \Phi'_i = 0$		$\nabla_{\Gamma}^2 \Phi''_i + \kappa_i''^2 \Phi''_i = 0$	equazione
$\Phi'_i(\vec{\rho}) = 0$	$\vec{\rho} \in C$	$\frac{\partial \Phi''_i(\vec{\rho})}{\partial n} = 0$	$\vec{\rho} \in C$ condizione al contorno
$\int_S \Phi_i'^2 ds = 1$		$\int_S \Phi_i''^2 ds = 1$	condizione di normalizzazione
$\int_S \Phi'_i \Phi'_j ds = 0$	$(i \neq j)$	$\int_S \Phi''_i \Phi''_j ds = 0$	$(i \neq j)$ ortogonalità
$\gamma_i' = \sqrt{\kappa_i'^2 - k^2}$		$\gamma_i'' = \sqrt{\kappa_i''^2 - k^2}$	costante di propagazione
$\vec{e}'_i = -\frac{\nabla_{\Gamma} \Phi'_i}{\kappa_i'}$		$\vec{h}''_i = -\frac{\nabla_{\Gamma} \Phi''_i}{\kappa_i''}$	} vettori modali
$\vec{h}'_i = \vec{u}_z \times \vec{e}'_i$		$\vec{e}''_i = -\vec{u}_z \times \vec{h}''_i$	
$Z'_i = \eta \frac{\gamma_i'}{jk}$		$Z''_i = \eta \frac{jk}{\gamma_i''}$	impedenza modale
$\vec{E}'_i = \vec{e}'_i V'_i - j \vec{u}_z \frac{\kappa_i'}{k} \Phi'_i \eta I'_i$		$\vec{E}''_i = \vec{e}''_i V''_i$	} campi modali
$\vec{H}'_i = \vec{h}'_i I'_i$		$\vec{H}''_i = \vec{h}''_i I''_i - j \vec{u}_z \frac{\kappa_i''}{k} \Phi''_i \frac{V''_i}{\eta}$	
<p>Si nota che la condizione di normalizzazione implica che i potenziali Φ_i, sia TE che TM, siano dimensionalmente uguali all'inverso di una lunghezza, così che risulta adimensionale il rapporto Φ_i/κ_i.</p>			

$$\omega < \omega_c \equiv \lambda > \lambda_c$$

$$\omega > \omega_c \equiv \lambda < \lambda_c$$

$$\alpha = \text{Re}(\gamma) = \frac{2\pi}{\lambda_c} \sqrt{1 - \frac{\lambda_c^2}{\lambda^2}} = \frac{\omega_c}{v} \sqrt{1 - \frac{\omega^2}{\omega_c^2}}$$

$$\beta = \text{Im}(\gamma) = 0$$

$$Z' = -j\eta \frac{\alpha}{k} = -j\eta \sqrt{\omega_c^2/\omega^2 - 1}$$

$$Z'' = j\eta \frac{k}{\alpha} = j \frac{\eta}{\sqrt{\omega_c^2/\omega^2 - 1}}$$

Z immaginarie

$$\alpha = \text{Re}(\gamma) = 0$$

$$\beta = \text{Im}(\gamma) = \frac{2\pi}{\lambda} \sqrt{1 - \frac{\lambda^2}{\lambda_c^2}} = \frac{\omega}{v} \sqrt{1 - \frac{\omega_c^2}{\omega^2}}$$

$$Z' = \eta \frac{\beta}{k} = \eta \frac{\lambda}{\lambda_g} = \eta \sqrt{1 - \omega_c^2/\omega^2} < \eta$$

$$Z'' = \eta \frac{k}{\beta} = \eta \frac{\lambda_g}{\lambda} = \frac{\eta}{\sqrt{1 - \omega_c^2/\omega^2}} > \eta$$

Z reali

$$\begin{aligned}\vec{E}(\vec{\rho}, z) &= \sum_i \left(\vec{e}'_i(\vec{\rho}) V'_i(z) - j \vec{u}_z \frac{\kappa'_i}{k} \Phi'_i(\vec{\rho}) \eta I'_i(z) \right) + \sum_i \vec{e}''_i(\vec{\rho}) V''_i(z) \\ \vec{H}(\vec{\rho}, z) &= \sum_i \vec{h}'_i(\vec{\rho}) I'_i(z) + \sum_i \left(\vec{h}''_i(\vec{\rho}) I''_i(z) - j \vec{u}_z \frac{\kappa''_i}{k} \Phi''_i(\vec{\rho}) \frac{V''_i(z)}{\eta} \right)\end{aligned}$$

$$\begin{aligned}V_i(z) &= V_i^+ e^{-\gamma_i z} + V_i^- e^{\gamma_i z} \\ I_i(z) &= \frac{V_i^+ e^{-\gamma_i z} - V_i^- e^{\gamma_i z}}{Z_i}\end{aligned}$$

$$\frac{1}{2} \int_S \operatorname{Re}(\vec{E} \times \vec{H}^*) \cdot \vec{u}_z \, ds = \frac{1}{2} \sum_i \operatorname{Re}(V_i' I_i'^*) + \frac{1}{2} \sum_i \operatorname{Re}(V_i'' I_i''^*)$$

$$\int_S \vec{e}_i \times \vec{h}_i \cdot \vec{u}_z \, ds = 1$$

$$\int_S \vec{e}_i \times \vec{h}_j \cdot \vec{u}_z \, ds = 0 \quad (i \neq j)$$

$$P_i = \frac{|V_i^+|^2}{2Z_i} - \frac{|V_i^-|^2}{2Z_i}$$

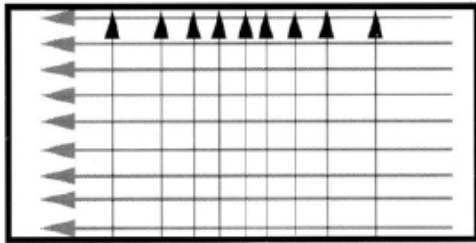
se il modo i si propaga
(Z_i reale)

$$P_i = \frac{\operatorname{Im}(V_i^+ V_i^{-*})}{\operatorname{Im}(Z_i)}$$

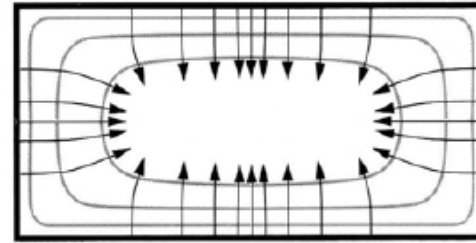
Se il modo i è evanescente
(Z_i immaginaria)

guida rettangolare

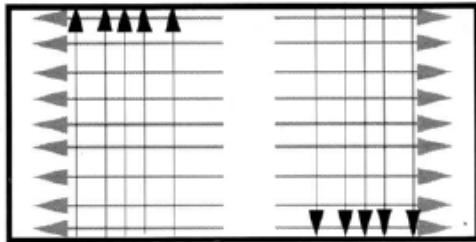
modi TM_{mn} ($m, n = 1, 2, \dots$)	modi TE_{mn} ($m, n = 0, 1, 2, \dots$ escluso $m = n = 0$)
$\kappa_{mn} = \pi \sqrt{(m/a)^2 + (n/b)^2}$ $\lambda_{mn} = \frac{2\pi}{\kappa_{mn}} = 2 / \sqrt{(m/a)^2 + (n/b)^2}$	
$\Phi'_{mn} = \frac{2}{\sqrt{ab}} S_x S_y$	$\Phi''_{mn} = \frac{2}{\sqrt{\chi_{mn} ab}} C_x C_y$
$\vec{e}'_{mn} = \frac{\lambda_{mn}}{\sqrt{ab}} (-\vec{u}_x \frac{m}{a} C_x S_y - \vec{u}_y \frac{n}{b} S_x C_y)$	$\vec{e}''_{mn} = \frac{\lambda_{mn}}{\sqrt{\chi_{mn} ab}} (\vec{u}_x \frac{n}{b} C_x S_y - \vec{u}_y \frac{m}{a} S_x C_y)$
$\vec{h}'_{mn} = \frac{\lambda_{mn}}{\sqrt{ab}} (\vec{u}_x \frac{n}{b} S_x C_y - \vec{u}_y \frac{m}{a} C_x S_y)$	$\vec{h}''_{mn} = \frac{\lambda_{mn}}{\sqrt{\chi_{mn} ab}} (\vec{u}_x \frac{m}{a} S_x C_y + \vec{u}_y \frac{n}{b} C_x S_y)$
$S_x = \sin \frac{m\pi x}{a} \quad S_y = \sin \frac{n\pi y}{b}$	$C_x = \cos \frac{m\pi x}{a} \quad C_y = \cos \frac{n\pi y}{b}$
$\chi_{mn} = \begin{cases} 2 & \text{se } m = 0 \text{ o } n = 0 \\ 1 & \text{se } m \neq 0 \text{ e } n \neq 0 \end{cases}$	



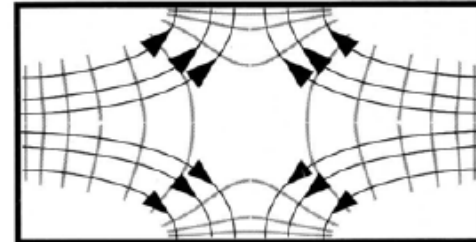
TE_{10}



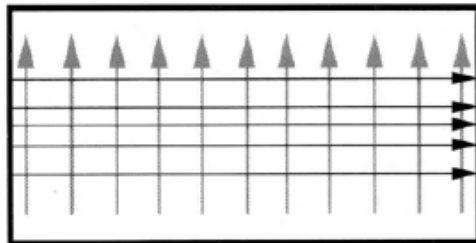
TM_{11}



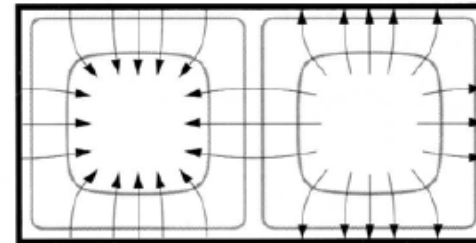
TE_{20}



TE_{11}



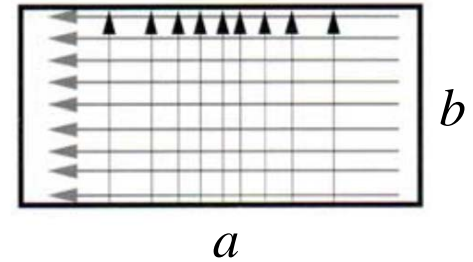
TE_{01}



TM_{21}

modo TE₁₀ in guida rettangolare

$$\kappa_{10}'' = \pi/a \quad \lambda_{10}'' = 2a \quad f_{10}'' = v/(2a)$$



$$\Phi_{10}'' = \sqrt{\frac{2}{ab}} \cos \frac{\pi x}{a} \quad \text{potenziale}$$

$$\vec{e}_{10}'' = -\sqrt{\frac{2}{ab}} \vec{u}_y \sin \frac{\pi x}{a} \quad \text{vettori modali}$$

$$\vec{h}_{10}'' = \sqrt{\frac{2}{ab}} \vec{u}_x \sin \frac{\pi x}{a}$$

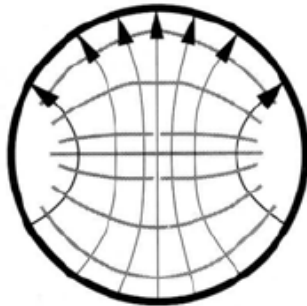
campi modali

$$\vec{E}_{10}'' = -\sqrt{\frac{2}{ab}} \vec{u}_y \sin \frac{\pi x}{a} V_{10}''(z)$$

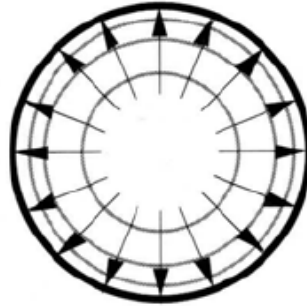
$$\vec{H}_{10}'' = \sqrt{\frac{2}{ab}} \left(\vec{u}_x \sin \frac{\pi x}{a} I_{10}''(z) - j \vec{u}_z \cos \frac{\pi x}{a} \frac{\lambda}{2a\eta} V_{10}''(z) \right)$$

guida d'onda circolare

modi $\text{TM}_{mn}^{(e)}$ ($m = 0, 1, 2, \dots; n = 1, 2, \dots$)		
$\kappa'_{mn} = \frac{x_{mn}}{a}$	x_{mn} è la n -esima radice dell'equazione $J_m(x) = 0$ (zero escluso)	$\lambda'_{mn} = \frac{2\pi a}{x_{mn}}$
$\Phi'_{mn} = F'_{mn} J_m(\kappa'_{mn} r) \cos m\varphi$		
$\vec{e}'_{mn} = -F'_{mn} \left(\vec{u}_r J'_m(\kappa'_{mn} r) \cos m\varphi - \vec{u}_\varphi \frac{m J_m(\kappa'_{mn} r)}{\kappa'_{mn} r} \sin m\varphi \right)$		
$\vec{h}'_{mn} = -F'_{mn} \left(\vec{u}_r \frac{m J_m(\kappa'_{mn} r)}{\kappa'_{mn} r} \sin m\varphi + \vec{u}_\varphi J'_m(\kappa'_{mn} r) \cos m\varphi \right)$		
$F'_{mn} = \frac{\chi_m}{a \sqrt{\pi} J_{m+1}(x_{mn})}$	$J'_m(\kappa'_{mn} r) = \left. \frac{dJ_m(x)}{dx} \right _{x=\kappa'_{mn} r}$	$\chi_m = \begin{cases} 1 & \text{se } m = 0 \\ \sqrt{2} & \text{se } m \neq 0 \end{cases}$
modi $\text{TE}_{mn}^{(e)}$ ($m = 0, 1, 2, \dots; n = 1, 2, \dots$)		
$\kappa''_{mn} = \frac{x'_{mn}}{a}$	x'_{mn} è la n -esima radice dell'equazione $J'_m(x) = 0$ (zero escluso)	$\lambda''_{mn} = \frac{2\pi a}{x'_{mn}}$
$\Phi''_{mn} = F''_{mn} J_m(\kappa''_{mn} r) \cos m\varphi$		
$\vec{e}''_{mn} = F''_{mn} \left(\vec{u}_r \frac{m J_m(\kappa''_{mn} r)}{\kappa''_{mn} r} \sin m\varphi + \vec{u}_\varphi J'_m(\kappa''_{mn} r) \cos m\varphi \right)$		
$\vec{h}''_{mn} = F''_{mn} \left(-\vec{u}_r J'_m(\kappa''_{mn} r) \cos m\varphi + \vec{u}_\varphi \frac{m J_m(\kappa''_{mn} r)}{\kappa''_{mn} r} \sin m\varphi \right)$		
$F''_{mn} = \frac{\chi \kappa''_{mn}}{J_m(x'_{mn}) \sqrt{\pi (x'^2_{mn} - m^2)}}$		
modi $\text{TM}_{mn}^{(o)}$ e $\text{TE}_{mn}^{(o)}$ ($m = 1, 2, \dots; n = 1, 2, \dots$)		
le espressioni dei potenziali e dei vettori modali relativi ai modi dispari si deducono dalle corrispondenti espressioni relative ai modi pari cambiando $\cos \rightarrow \sin$ e $\sin \rightarrow -\cos$.		



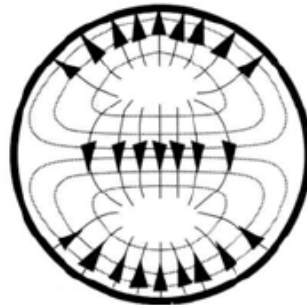
TE₁₁
 $(\lambda''_{11} = 3.412 a)$



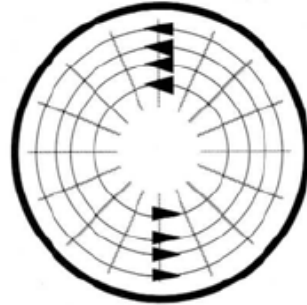
TM₀₁
 $(\lambda'_{01} = 2.613 a)$



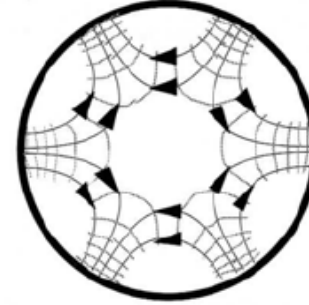
TE₂₁
 $(\lambda''_{21} = 2.057 a)$



TM₁₁
 $(\lambda'_{11} = 1.640 a)$



TE₀₁
 $(\lambda''_{01} = 1.640 a)$



TE₃₁
 $(\lambda''_{31} = 1.496 a)$

modo TE₁₁ in guida circolare

modo TE₁₁^(e)

$$\kappa_{11}'' = x'_{11} / a \approx 1.841 / a$$

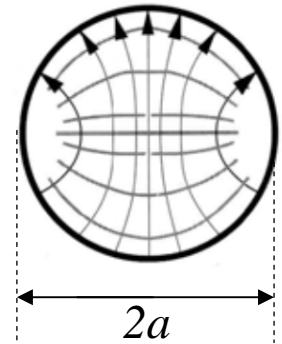
$$\lambda_{11}'' = 2\pi a / x'_{11} \approx 3.412 a$$

$$\Phi_{11}''^{(e)} = F_{11}'' J_1(\kappa_{11}'' r) \cos \varphi$$

$$\vec{e}_{11}''^{(e)} = F_{11}'' \left(\vec{u}_r \frac{J_1(\kappa_{11}'' r)}{\kappa_{11}'' r} \sin \varphi + \vec{u}_\varphi J_1'(\kappa_{11}'' r) \cos \varphi \right)$$

$$\vec{h}_{11}''^{(e)} = F_{11}'' \left(-\vec{u}_r J_1'(\kappa_{11}'' r) \cos \varphi + \vec{u}_\varphi \frac{J_1(\kappa_{11}'' r)}{\kappa_{11}'' r} \sin \varphi \right)$$

$$F_{11}'' = \frac{\sqrt{2} \kappa_{11}''}{J_1(x'_{11}) \sqrt{\pi (x'_{11})^2 - 1}} \approx \frac{1.633}{a}$$

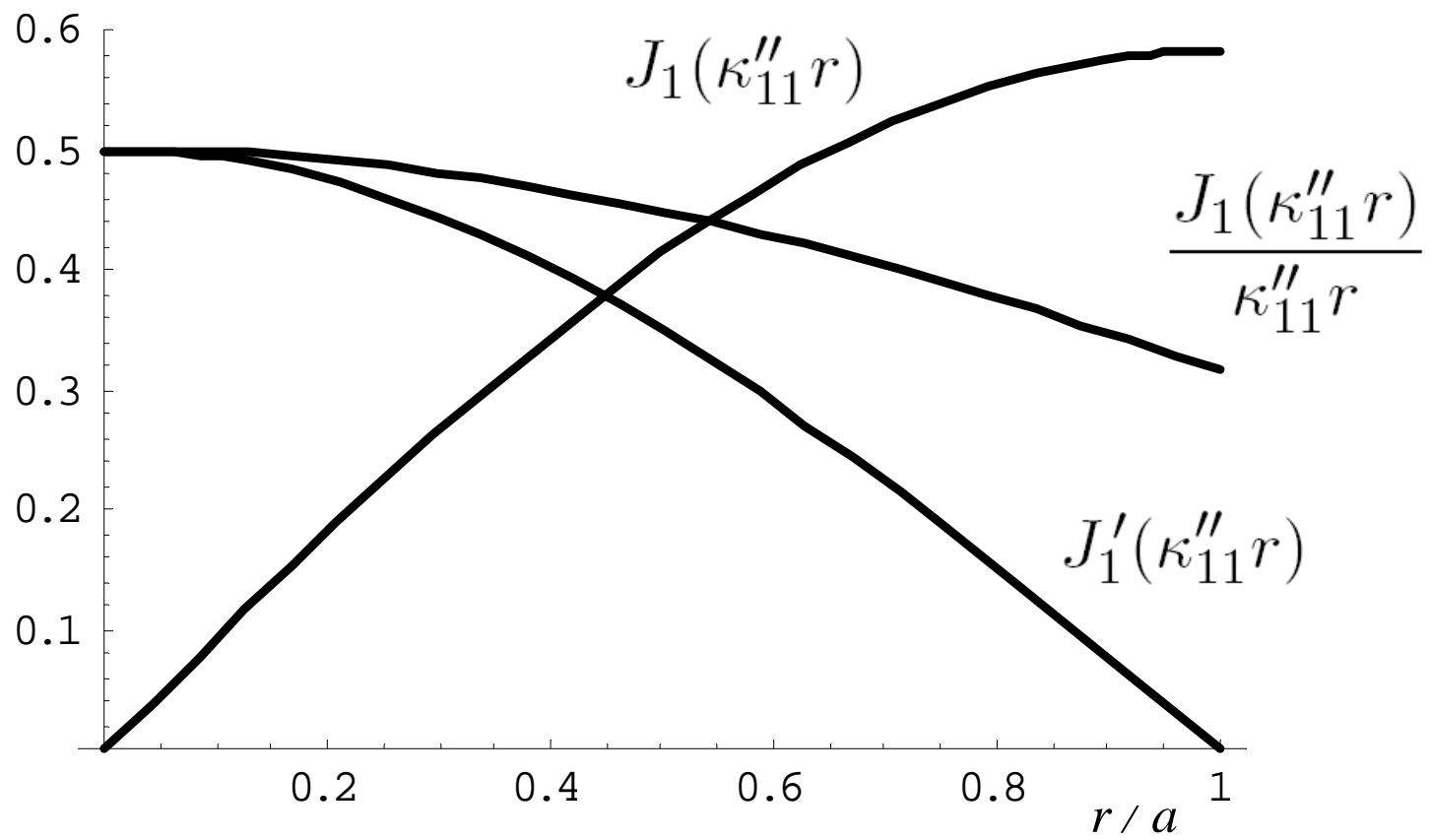


le espressioni del potenziale e dei vettori modali relativi al modo dispari si deducono dalle precedenti cambiando $\cos \varphi \rightarrow \sin \varphi$ e $\sin \varphi \rightarrow -\cos \varphi$.

campi modali

$$\vec{E}_{11}'' = \vec{e}_{11}''^{(e)} V_{11}''^{(e)}(z) + \vec{e}_{11}''^{(o)} V_{11}''^{(o)}(z)$$

$$\vec{H}_{11}'' = \vec{h}_{11}''^{(e)} I_{11}''^{(e)}(z) + \vec{h}_{11}''^{(o)} I_{11}''^{(o)}(z) - j \vec{u}_z \frac{\lambda}{\lambda_{11}''} \left(\Phi_{11}''^{(e)} \frac{V_{11}''^{(e)}(z)}{\eta} + \Phi_{11}''^{(o)} \frac{V_{11}''^{(o)}(z)}{\eta} \right)$$



modi TEM

$$\nabla_{\text{T}}^2 \Phi_i^{\circ} = 0$$

equazione

$$\Phi_i^{\circ}(\vec{\rho}) = 0$$

$$\vec{\rho} \in C_0$$

$$\Phi_i^{\circ}(\vec{\rho}) = \text{cost}_i^{(n)}$$

$$\vec{\rho} \in C_n$$

condizione al contorno

$$\gamma_i^{\circ} = jk$$

costante di propagazione

$$\vec{e}_i^{\circ} = -\nabla_{\text{T}} \Phi_i^{\circ}$$

vettori modali

$$\vec{h}_i^{\circ} = \vec{u}_z \times \vec{e}_i^{\circ}$$

$$Z_i^{\circ} = \eta$$

impedenza modale

$$\vec{E}_i^{\circ} = \vec{e}_i^{\circ} V_i^{\circ}$$

campi modali

$$\vec{H}_i^{\circ} = \vec{h}_i^{\circ} I_i^{\circ}$$

$$\begin{aligned}
V_i^{(n)} &= \int_{\vec{\rho}_0}^{\vec{\rho}_n} \vec{E}_i^{\circ} \cdot d\vec{\ell} = -V_i^{\circ} \int_{\vec{\rho}_0}^{\vec{\rho}_n} \nabla_{\text{T}} \Phi_i^{\circ}(\vec{\rho}) \cdot d\vec{\ell} = V_i^{\circ} \left(\Phi_i^{\circ}(\vec{\rho}_n) - \Phi_i^{\circ}(\vec{\rho}_0) \right) \\
&= V_i^{\circ} \text{cost}_i^{(n)}
\end{aligned}$$

$$\vec{J}_s = -\vec{n} \times \vec{H}_i^{\circ} = I_i^{\circ} \vec{n} \times (\vec{u}_z \times \nabla_{\text{T}} \Phi_i^{\circ}) = \vec{u}_z I_i^{\circ} \frac{\partial \Phi_i^{\circ}}{\partial n}$$

$$I_i^{(n)} = I_i^{\circ} \int_{C_n} \frac{\partial \Phi_i^{\circ}}{\partial n} dc$$

$$Z_i^{(n)} = \eta \text{cost}_i^{(n)} \left(\int_{C_n} \frac{\partial \Phi_i^{\circ}}{\partial n} dc \right)^{-1} \quad \text{impedenza caratteristica}$$

$$\begin{aligned}
P_i^{\circ} &= \frac{1}{2} \operatorname{Re} \left(\int_S \vec{E}_i^{\circ} \times \vec{H}_i^{\circ*} \cdot \vec{u}_z \, ds \right) \\
&= \frac{1}{2} \operatorname{Re} \left(V_i^{\circ} I_i^{\circ*} \right) \int_S \vec{e}_i^{\circ} \times \vec{h}_i^{\circ} \cdot \vec{u}_z \, ds
\end{aligned}$$

$$\begin{aligned}
P_i^{\circ} &= \frac{1}{2} \operatorname{Re} \left(\sum_{n=1}^N \left(V_i^{\circ} \operatorname{cost}_i^{(n)} \right) \left(I_i^{\circ} \int_{C_n} \frac{\partial \Phi_i^{\circ}}{\partial n} \, dc \right)^* \right) \\
&= \frac{1}{2} \sum_{n=1}^N \operatorname{Re} \left(V_i^{(n)} I_i^{(n)*} \right)
\end{aligned}$$

$$\int_S \vec{e}_i^{\circ} \times \vec{h}_j^{\circ} \cdot \vec{u}_z \, ds = \begin{cases} \sum_{n=1}^N \operatorname{cost}_i^{(n)} \int_{C_n} \frac{\partial \Phi_j^{\circ}}{\partial n} \, dc & i = j \\ \mathbf{0} & i \neq j \end{cases}$$

modo TEM (un solo conduttore interno)

$$\nabla_{\mathbf{T}}^2 \Phi^{\circ} = 0$$

equazione

$$\Phi^{\circ}(\vec{\rho}) = 0$$

$$\vec{\rho} \in C_0$$

$$\Phi^{\circ}(\vec{\rho}) = 1$$

$$\vec{\rho} \in C_1$$

condizione al contorno

$$\gamma^{\circ} = jk$$

costante di propagazione

$$\vec{e}^{\circ} = -\nabla_{\mathbf{T}} \Phi^{\circ}$$

$$\vec{h}^{\circ} = \frac{Z_c}{\eta} \vec{u}_z \times \vec{e}^{\circ}$$

vettori modali

$$Z_c = \eta \left(\int_{C_1} \frac{\partial \Phi^{\circ}}{\partial n} dc \right)^{-1}$$

impedenza caratteristica

$$\vec{E}^{\circ} = \vec{e}^{\circ} V$$

$$\vec{H}^{\circ} = \vec{h}^{\circ} I$$

campi modali

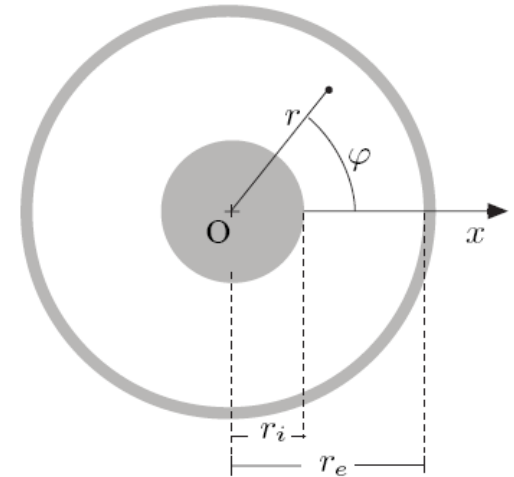
$$\int_S \vec{e}^{\circ} \cdot \vec{e}^{\circ} ds = \frac{\eta}{Z_c}$$

$$\int_S \vec{h}^{\circ} \cdot \vec{h}^{\circ} ds = \frac{Z_c}{\eta}$$

$$\int_S \vec{e}^{\circ} \times \vec{h}^{\circ} \cdot \vec{u}_z ds = 1$$

modo TEM del cavo coassiale

$$\begin{aligned}\Phi^{\circ} &= \frac{\ln(r_e/r)}{\ln(r_e/r_i)} \\ \vec{e}^{\circ} &= \frac{1}{r \ln(r_e/r_i)} \vec{u}_r \\ \vec{h}^{\circ} &= \frac{Z_c}{\eta} \vec{u}_z \times \vec{e}^{\circ} = \frac{1}{2\pi r} \vec{u}_\varphi \\ Z_c &= \frac{\eta}{2\pi} \ln(r_e/r_i)\end{aligned}$$

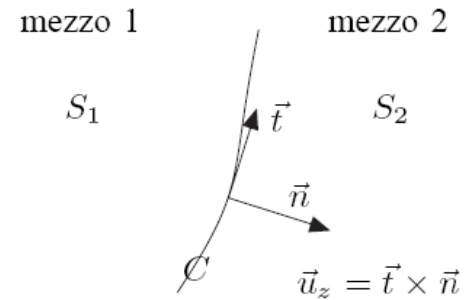


$$\vec{E}^{\circ}(r, \varphi, z) = \vec{e}^{\circ}(r, \varphi) V(z)$$

$$\vec{H}^{\circ}(r, \varphi, z) = \vec{h}^{\circ}(r, \varphi) I(z)$$

il primo modo superiore è il TE_{11} : $\lambda_{11}'' \approx \pi (r_i + r_e)$

condizioni su un'interfaccia



$$\nabla_{\Gamma}^2 \Phi_1(\vec{\rho}) + \kappa_1^2 \Phi_1(\vec{\rho}) = 0 \quad \vec{\rho} \in S_1$$

$$\kappa_1^2 = k_1^2 + \gamma^2$$

$$\nabla_{\Gamma}^2 \Phi_2(\vec{\rho}) + \kappa_2^2 \Phi_2(\vec{\rho}) = 0 \quad \vec{\rho} \in S_2$$

$$\kappa_2^2 = k_2^2 + \gamma^2$$

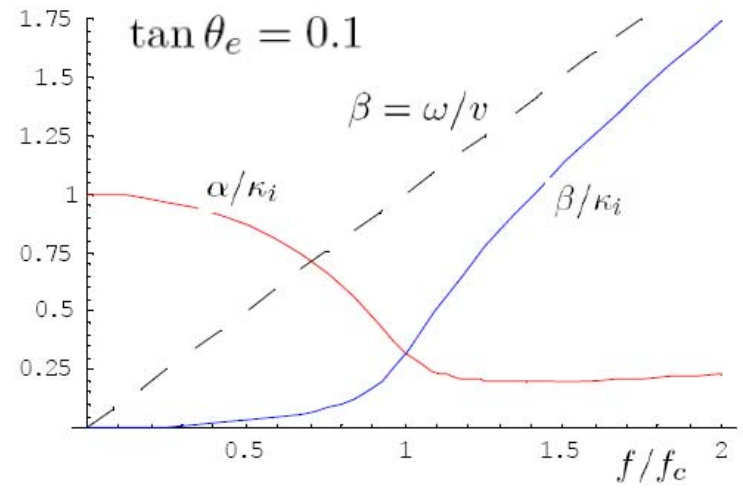
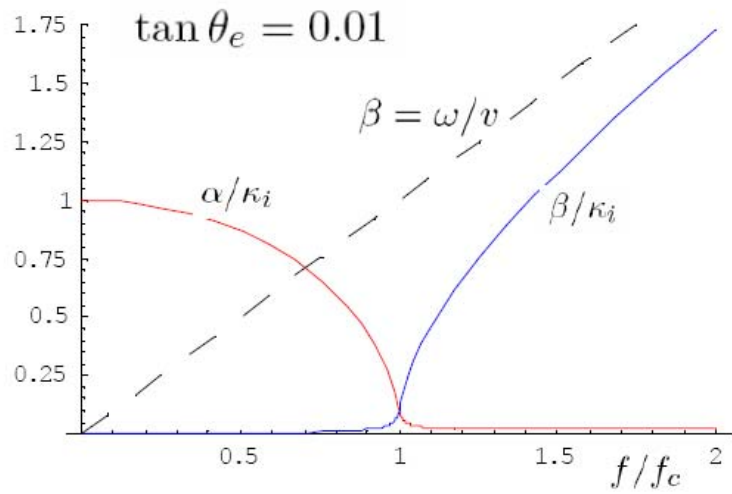
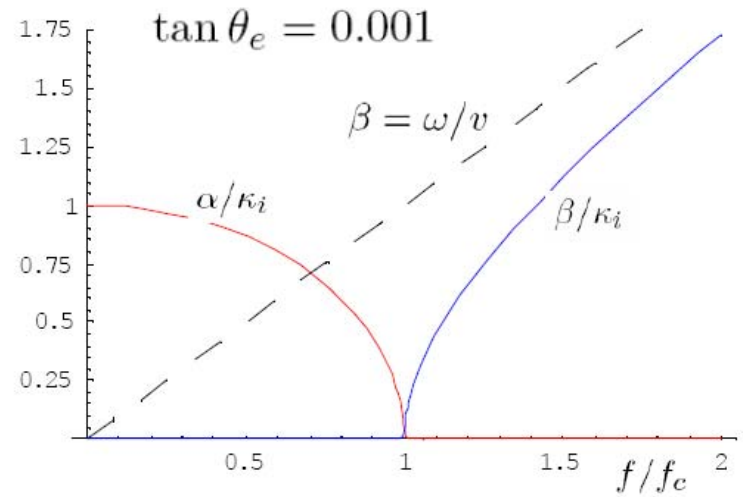
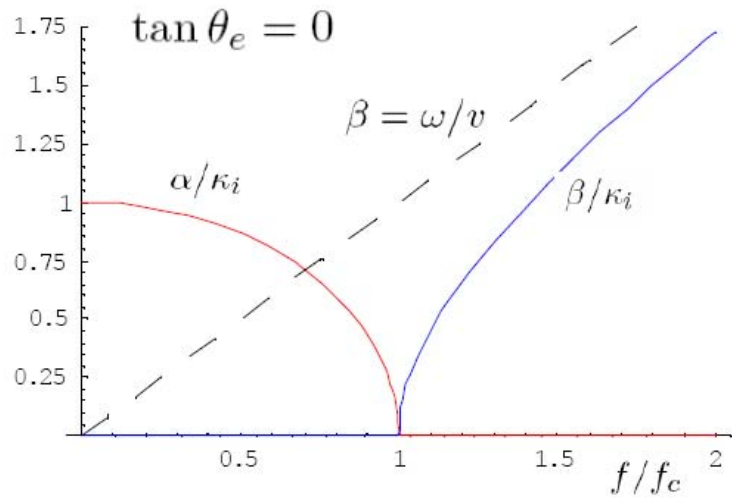
$$\left. \begin{aligned} \frac{\eta_1 \gamma}{j k_1 Z} \frac{\partial \Phi_1'(\vec{\rho})}{\partial t} + \frac{\partial \Phi_1''(\vec{\rho})}{\partial n} &= \frac{\eta_2 \gamma}{j k_2 Z} \frac{\partial \Phi_2'(\vec{\rho})}{\partial t} + \frac{\partial \Phi_2''(\vec{\rho})}{\partial n} \end{aligned} \right\}$$

$$\left. \begin{aligned} \frac{\gamma Z}{j k_1 \eta_1} \frac{\partial \Phi_1''(\vec{\rho})}{\partial t} - \frac{\partial \Phi_1'(\vec{\rho})}{\partial n} &= \frac{\gamma Z}{j k_2 \eta_2} \frac{\partial \Phi_2''(\vec{\rho})}{\partial t} - \frac{\partial \Phi_2'(\vec{\rho})}{\partial n} \end{aligned} \right\}$$

$$\left. \begin{aligned} \frac{\eta_1 \kappa_1^2}{k_1} \Phi_1'(\vec{\rho}) &= \frac{\eta_2 \kappa_2^2}{k_2} \Phi_2'(\vec{\rho}) \end{aligned} \right\}$$

$$\left. \begin{aligned} \frac{\kappa_1^2}{k_1 \eta_1} \Phi_1''(\vec{\rho}) &= \frac{\kappa_2^2}{k_2 \eta_2} \Phi_2''(\vec{\rho}) \end{aligned} \right\}$$

$\vec{\rho} \in C$



linea bifilare

$$\Phi^{\circ} = \frac{1}{F} \ln \frac{(x - x_0)^2 + y^2}{(x + x_0)^2 + y^2}$$

$$\vec{e}^{\circ} = \frac{4x_0(x^2 - x_0^2 - y^2)\vec{u}_x + 2xy\vec{u}_y}{F ((x - x_0)^2 + y^2)((x + x_0)^2 + y^2)}$$

$$\vec{h}^{\circ} = \frac{Z_c}{\eta} \frac{-2xy\vec{u}_x + 4x_0(x^2 - x_0^2 - y^2)\vec{u}_y}{F ((x - x_0)^2 + y^2)((x + x_0)^2 + y^2)}$$

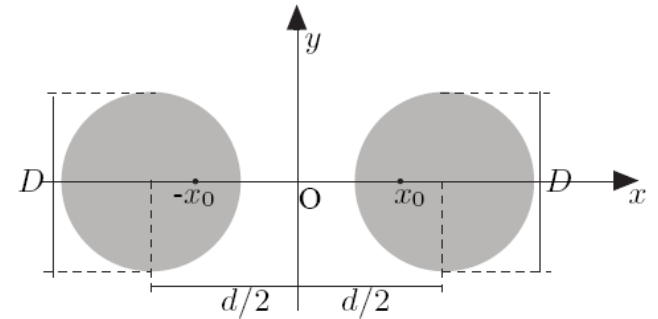
$$Z_c = \frac{\eta}{\pi} \operatorname{arccosh} \frac{d}{D}$$

$$\vec{E}^{\circ} = \vec{e}^{\circ} V^{\circ}(z)$$

$$\vec{H}^{\circ} = \vec{h}^{\circ} I^{\circ}(z)$$

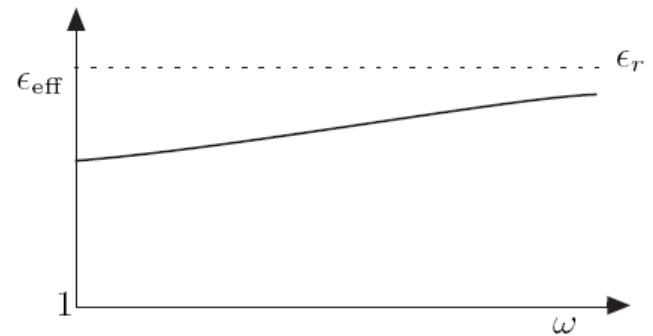
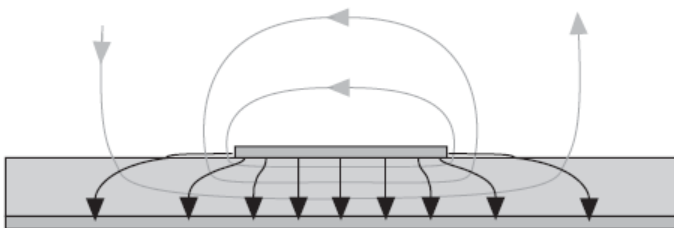
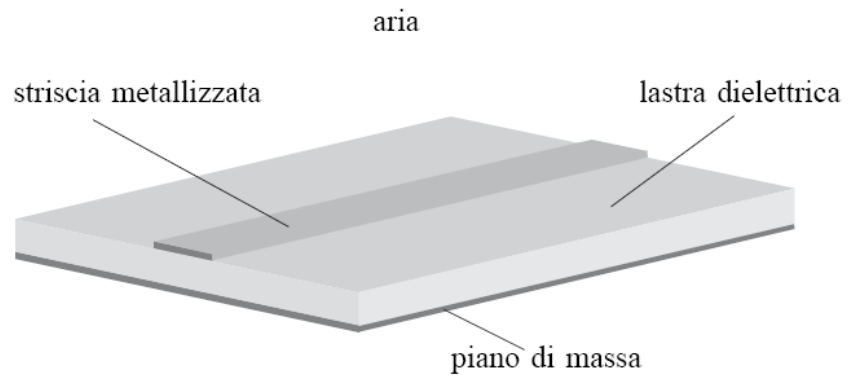
$$x_0 = \frac{1}{2} \sqrt{d^2 - D^2}$$

$$F = \ln \frac{d - 2x_0}{d + 2x_0}$$

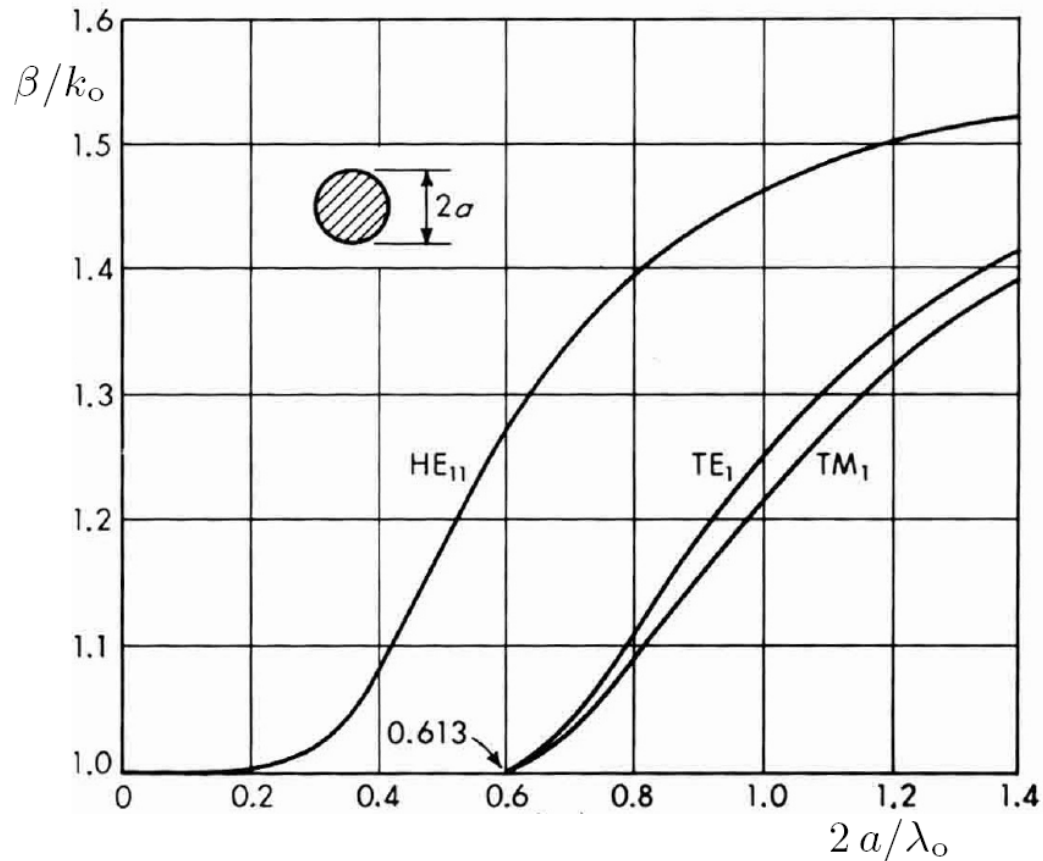


linea in microstriscia

$$\epsilon_{\text{eff}} = \frac{\beta^2}{(\omega/c)^2}$$



guida dielettrica



diagrammi di dispersione dei primi modi confinati di una guida dielettrica a sezione circolare in polistirene ($\epsilon_r = 2.56$).

lastra dielettrica



$$0 \leq x < d/2$$

$$\frac{d^2 \Phi''(x)}{d x^2} + \underbrace{(\omega/c)^2 n^2 - \beta^2}_{k_x^2} \Phi''(x) = 0$$

$$d/2 < x$$

$$\frac{d^2 \Phi''(x)}{d x^2} + \underbrace{(\omega/c)^2 - \beta^2}_{-\alpha^2} \Phi''(x) = 0$$

$$\lim_{x \rightarrow \infty} \Phi''(x) = 0$$

$$\lim_{x \rightarrow d/2^-} \frac{d \Phi''(x)}{d x} = \lim_{x \rightarrow d/2^+} \frac{d \Phi''(x)}{d x}$$

$$\lim_{x \rightarrow d/2^-} k_x^2 \Phi''(x) = \lim_{x \rightarrow d/2^+} -\alpha^2 \Phi''(x)$$

$$0 \leq x < d/2$$

$$d/2 < x$$

$$\Phi''(x) = B \sin(k_x x)$$

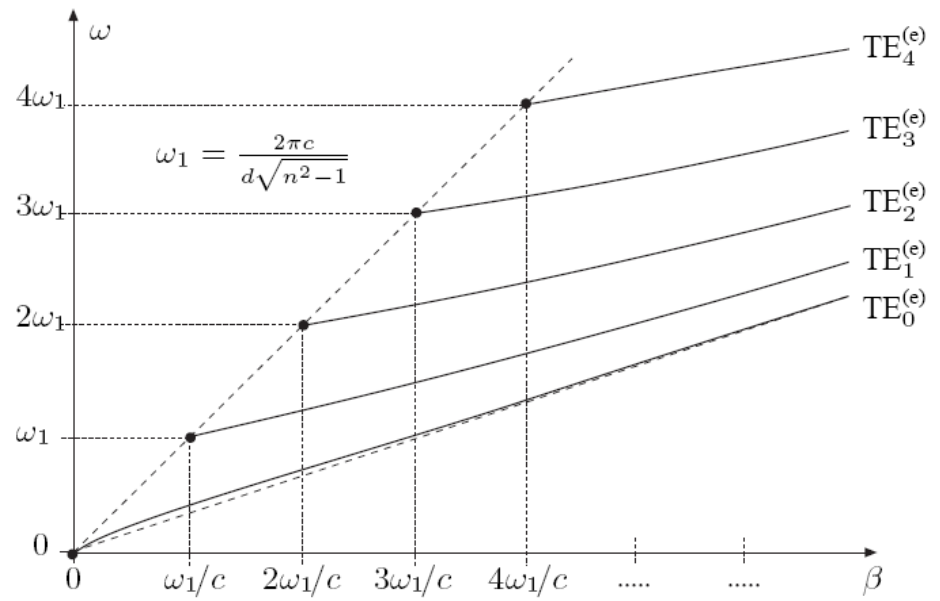
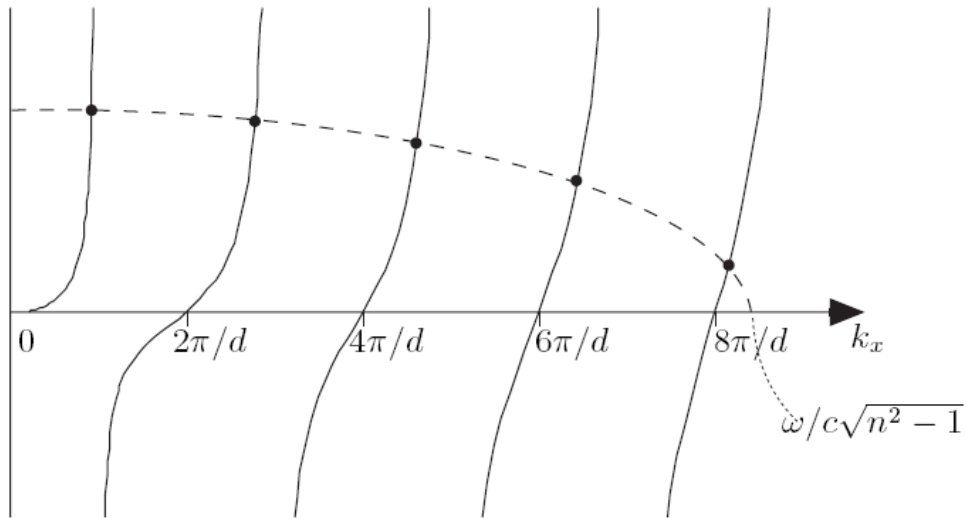
$$\Phi''(x) = A e^{-\alpha x}$$

$$\operatorname{Re}(\alpha) > 0$$

$$B k_x \cos k_x d/2 = -\alpha A e^{-\alpha d/2}$$

$$B k_x^2 \sin k_x d/2 = -\alpha^2 A e^{-\alpha d/2}$$

$$k_x \tan k_x d/2 = \underbrace{\sqrt{(\omega/c)^2 (n^2 - 1) - k_x^2}}_{\alpha}$$

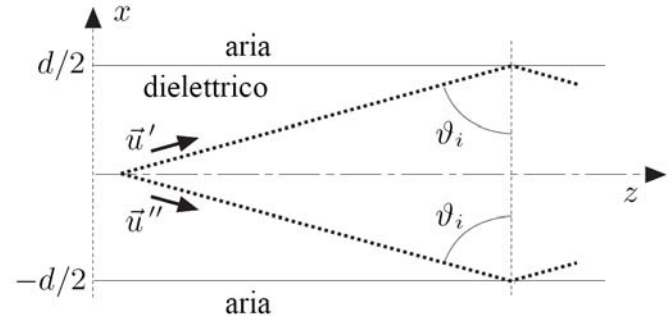


$$\vec{E} = \underbrace{\vec{u}_y \frac{A k_x}{2} e^{-jk_o n \vec{u}' \cdot \vec{r}}}_{\vec{E}'} + \underbrace{\vec{u}_y \frac{A k_x}{2} e^{-jk_o n \vec{u}'' \cdot \vec{r}}}_{\vec{E}''}$$

$$\vec{H} = \underbrace{\frac{n}{\eta_o} \vec{u}' \times \vec{E}'}_{\vec{H}'} + \underbrace{\frac{n}{\eta_o} \vec{u}'' \times \vec{E}''}_{\vec{H}''}$$

$$\vec{u}' = \frac{k_x \vec{u}_x + \beta \vec{u}_z}{k_o n}$$

$$\vec{u}'' = \frac{-k_x \vec{u}_x + \beta \vec{u}_z}{k_o n}$$



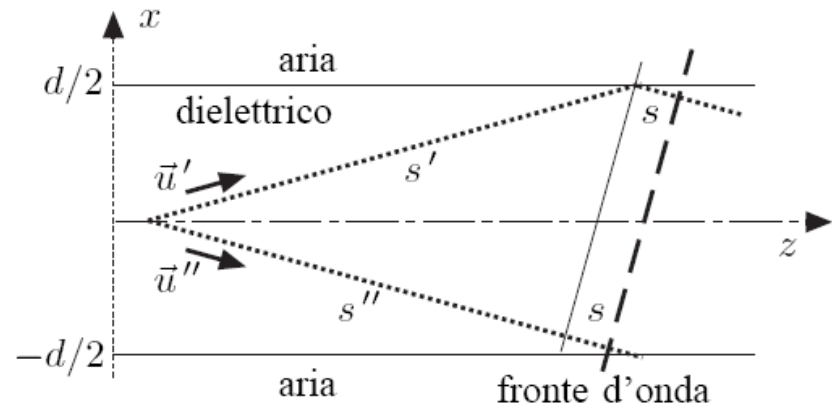
$$\tan \vartheta_i = \frac{\beta}{k_x} = \frac{\sqrt{1 + n^2 \tan^2(k_x d/2)}}{\sqrt{n^2 - 1}} > \frac{1}{\sqrt{n^2 - 1}} = \tan \vartheta_L$$

$$\beta \sqrt{n^2 - 1} = \sqrt{k_x^2 + n^2 \alpha^2} = k_x \sqrt{1 + n^2 \tan^2 k_x d/2}$$

$$\sin \vartheta_L = 1/n$$

$$\sin \vartheta_i = \beta / (k_o n)$$

$$\cos \vartheta_i = k_x / (k_o n)$$



$$-k_o n s' + \arg(\Gamma_{\perp}) - k_o n s = -k_o n (s'' + s)$$

$$\arg(\Gamma_{\perp}) = 2 \arctan \left(\sqrt{\sin^2 \vartheta_i - \sin^2 \vartheta_L} / \cos \vartheta_i \right)$$

$$s' = d/2 \sqrt{1 + \tan^2 \vartheta_i} = \frac{k_o n d}{2 k_x}$$

$$s'' = s' \cos(2(\pi/2 - \vartheta_i)) = -s' \cos(2\vartheta_i) = s'(1 - 2 \cos^2 \vartheta_i)$$

attenuazione nelle guide reali

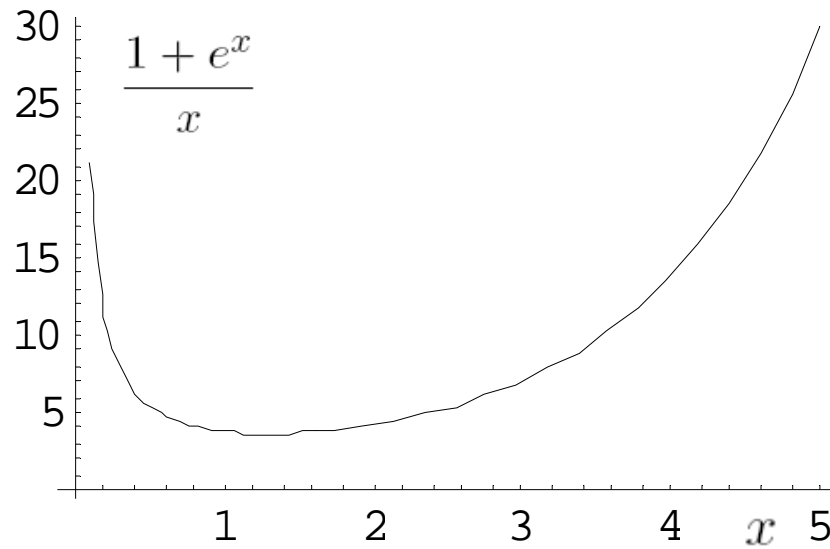
$$\begin{aligned}\alpha &= \frac{1}{2P} \left(\frac{dP_c}{dz} + \frac{dP_d}{dz} \right) \\ &= \frac{\int_C R_s |\vec{H}|^2 dc + \omega \int_S (\epsilon \tan \theta_e |E|^2 + \mu \tan \theta_m |H|^2) ds}{2 \int_S \operatorname{Re} (\vec{E} \times \vec{H}^*) \cdot \vec{u}_z ds}\end{aligned}$$

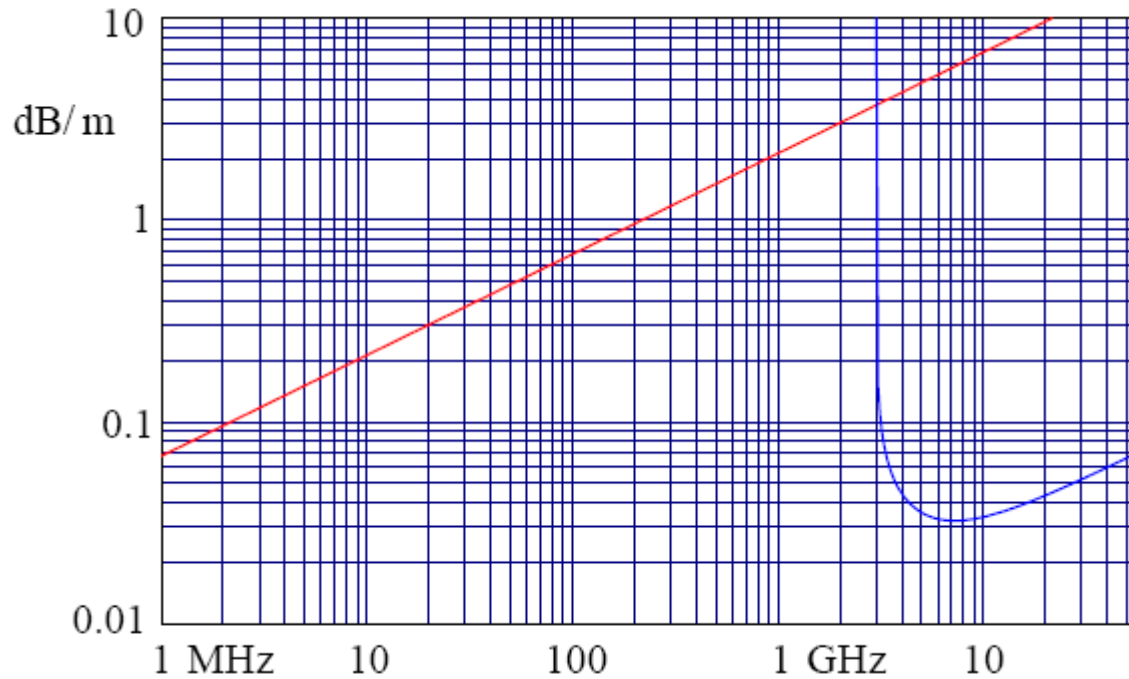
$$dP_c = \frac{1}{2} dz \int_C R_s |\vec{J}_s|^2 dc$$

$$dP_d = \frac{1}{2} \omega dz \int_S (\epsilon \tan \theta_e |E|^2 + \mu \tan \theta_m |H|^2) ds$$

$$2\alpha_c r_e \eta / R_s = \frac{1 + e^x}{x}$$

$$x = 2\pi Z_c / \eta = \ln(r_e / r_i)$$





Andamento dell'attenuazione del modo TEM di un tipico cavo coassiale e del modo TE₁₀ di una guida rettangolare in aria

